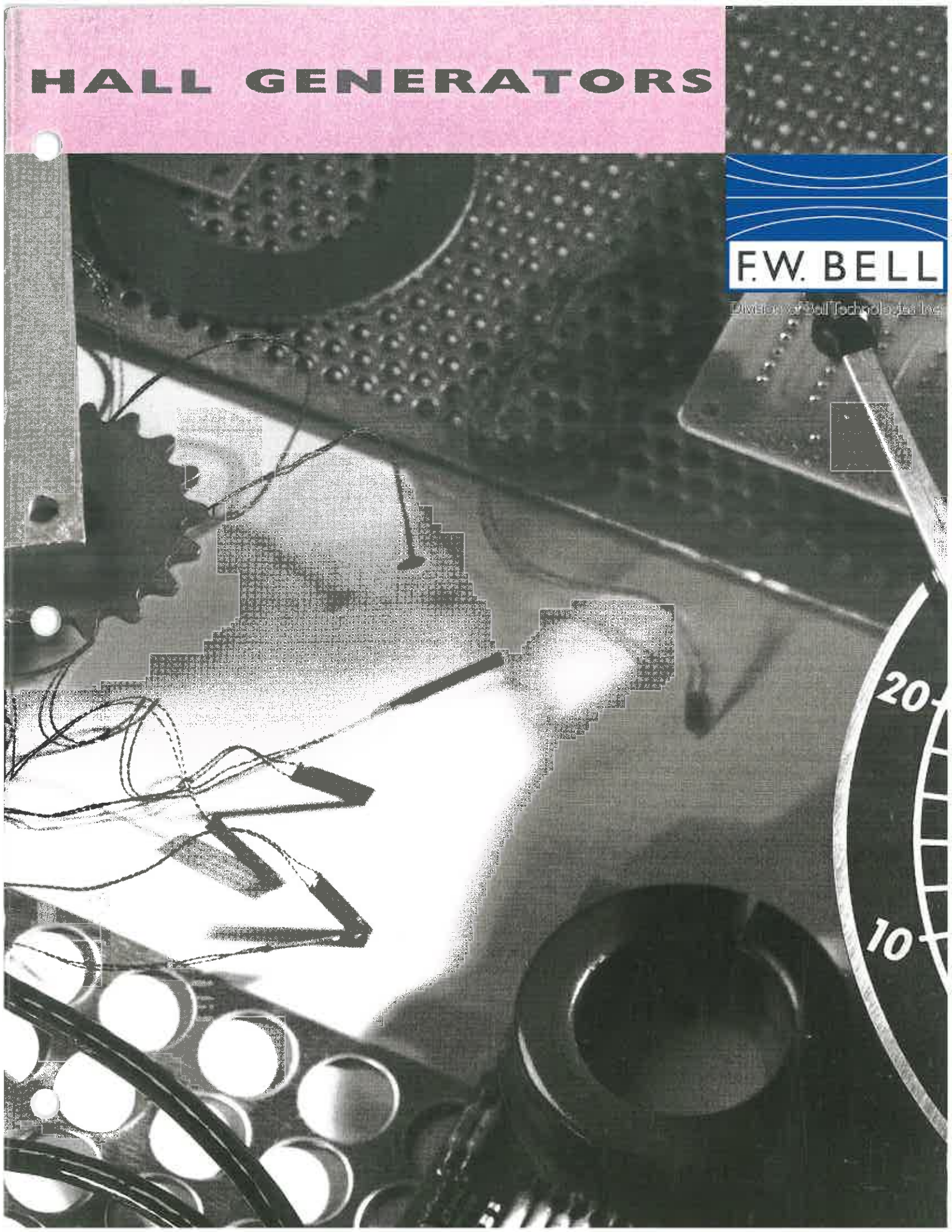


# HALL GENERATORS

**F.W. BELL**

Division of Bell Technologies Inc.



# AN INTRODUCTION TO THE HALL EFFECT

## Early Magnetism

Far back in the recesses of time man discovered the loadstone (Lodestone) in Magnesia in the district of Thessaly, Greece (hence the word "magnet").

They soon learned about its tendency to orient itself with the pole star, and used it as a guide to and from far away lands.

The first manufactured permanent magnets were needles that had been rubbed on the loadstone. The first electro-magnet was constructed in 1825, and the first scientific treatise in the nature of magnets was published 20 years later.

By the time that first electromagnet was built, permanent magnets had been in use for a variety of experimental and practical uses. Their strength was classified by the ratio of their weight to their lifting strength.



Photograph of Dr. Edwin Herbert Hall, 1887, courtesy of the Harvard University Archives

## The Discovery of "Hall Effect"

Edwin Herbert Hall discovered the "Hall effect" in 1879 while working on his doctoral thesis in Physics under the supervision of Professor Henry A. Rollin.<sup>1</sup> Dr. Hall was pursuing the question as to whether the resistance of a coil excited by a current was affected by the presence of a magnet. Through a

myriad of experiments and failures, Hall discovered that a magnetic field would skew equipotential lines in a current-carrying conductor. This effect is observed as a voltage (Hall voltage,  $V_H$ ) perpendicular to the direction of current in the conductor.

Hall conducted an experiment by putting a thin gold leaf on a glass plate and then tapping off the gold leaf at points down its length. He then conducted other experiments using various materials in place of the gold leaf, and various experimental placements of tapping points. In 1880, full details of Hall's experimentation with this phenomenon formed his doctoral thesis and was published in the *American Journal of Science* and in the *Philosophical Magazine*.<sup>2</sup>

Kelvin, himself a most distinguished scientist, called Hall's discovery comparable to the greatest ever made by Michael Faraday. The magnitude of this discovery is even more impressive considering how little was known about electricity in Hall's time. The electron, for instance, was not identified until more than 10 years later.<sup>3</sup>

## The Theory of the Hall Effect

The action of the Hall effect in a semi-conducting medium is adequately explained by quantum physics. However, in spite of its shortcomings, the classical approach is chosen here for its brevity.

A particle with charge  $Q$ , velocity,  $\vec{V}$ , and moving within a magnetic field,  $\vec{B}$ , will experience the Lorentz force,  $F=Q(\vec{V} \times \vec{B})$ . The force direction is mutually perpendicular to the directions of the particle velocity and the magnetic field. If a long, flat current-carrying conductor is placed in a magnetic field,

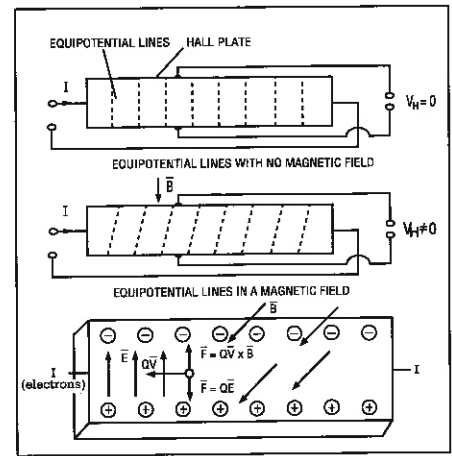




the moving charges will experience a net force mutually perpendicular to the direction of the current flow (longitudinal conductor axis) and the magnetic field. Under the influence of this force, the electrons will "pile up" on one edge of the conductor, and positive charges will gather on the other edge. An uneven lateral charge distribution results and gives rise to an electric field,  $\vec{E}$ , which exerts a force,  $\vec{F} = Q\vec{E}$ , opposite in direction to the Lorentz force. At equilibrium, the resultant forces balance (Fig. 2). This field, superimposed on the  $\vec{E}$  in the direction of the current flow, yields the skewed equipotential lines first noted by Hall (Fig. 1). The relation between the voltage, current, and magnetic field can be generalized as follows:

$$V_H = \gamma IB$$

$V_H$  = Hall voltage  
 $\gamma$  = a constant product sensitivity  
 $I$  = Hall current  
 $B$  = magnetic field perpendicular to Hall plate surface



**Figure 1**  
Explanation of the Hall effect.

This equation ignores many low level effects but will suffice for the depth of this discussion.

*Note: All  $\vec{B}$  fields in the article refer to the component of the external  $\vec{B}$  field that is normal to the surface of the Hall plate. A more general equation for Hall voltage is  $V_H = \gamma IB \cos \theta$ , where  $\theta$  is the angle between  $B$  and the normal to the Hall plate surface.*

## THE HALL GENERATOR

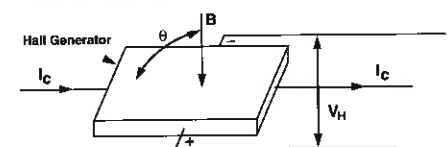
### From Theory To Practice

The "Hall effect" remained a laboratory curiosity until the latter half of this century because materials available prior to recent years only produced low levels of Hall voltage. With the advent of semiconductor technology and the development of various III-V compounds, it became possible to produce Hall voltages many orders of magnitude greater than with earlier materials. Thus, semiconductor technology launched the practical design and production of the Hall generator.

### What Is A Hall Generator

A Hall generator is a four-terminal, solid-state device capable of producing an output voltage,  $V_H$ , proportional to the

product of the input current,  $I_C$ , the magnetic flux density,  $B$ , and the sine of the angle between  $B$  and the plane of the Hall generator.



$$V_H = K_{HOC} I_C B \sin \theta \text{ or if } \sin \theta = 1 \text{ (i.e., } \theta = 90^\circ \text{)}$$

$$V_H = K_{HOC} I_C B \text{ or } V_H = \gamma_B B$$

where:  $V_H$  = Hall output voltage, mV  
 $K_{HOC} = \gamma_B$  (open circuit product sensitivity constant), mV/mA kG  
 $\gamma_B$  = magnetic sensitivity (loaded or unloaded) at a specified control current, mV/kG  
 $I_C$  = control current, mA (ac or dc)  
 $B$  = magnetic flux density, kG (ac or dc)

**Figure 2**

A reversal in the direction of either the magnetic field or the control current will result in a polarity change of  $V_H$ . A reversal in the direction of both will keep

1. C.L.Chin and C.R.Westgate (Editors), "The Hall Effect and Its Applications," Plenum Press, New York, 1979, p. 535.  
 2. Ibid., p. 523  
 3. Charles Coulestone Gillespie (Editor), "Dictionary of Scientific Bibliography," Charles Scribner's Sons, New York, 1970, p. 51.

See MIL-STD-793-1 (WP) for definitions

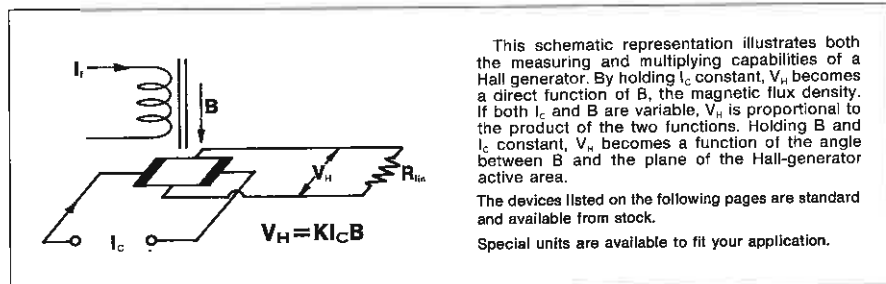


Figure 3

the polarity the same. By holding the control current constant, the Hall voltage may be used to measure magnetic flux density. Multiplication may be accomplished by varying both the control current and the magnetic field.

## Materials

The Hall effect is basically a majority carrier mechanism depending on the bulk-material properties of the semiconductor material. Unlike transistors and diodes, it is completely independent of surface effects, junction leakage currents and junction threshold voltages. These factors account for its high stability, reproducibility and reliability when compared to other semiconductor devices.

To obtain a high output voltage the active element must have a high Hall coefficient,  $R_H$ . Also, since the output is proportional to the current density through the element, its resistance should be as low as practical to prevent excessive heating. Since the noise output is essentially thermal<sup>4,5</sup> low resistance is also an important requirement for devices to be used at very low signal levels. Some of the semiconductor materials used for Hall generators are indium antimonide (InSb), indium arsenide (InAs) and gallium arsenide (GaAs). GaAs generators have high output and very high resistance making them relatively noisy and the temperature coefficient of the output voltage is less than  $-0.1\%/^{\circ}\text{C}$ . InSb has high output and low resistance, but the temperature coefficient of the output voltage is about  $-1\%/^{\circ}\text{C}$ . InAs has less output than InSb, but its temperature

coefficient is less than  $-0.1\%/^{\circ}\text{C}$  and its resistance is also low. These considerations make InAs the most suitable material for many Hall effect applications.

InAs Hall generators may also be made of deposited thin films. These units do not exhibit the same low resistance and high mobilities as their bulk-material counterparts, but they do offer advantages which may be realized in many applications. These advantages include lower current requirements for comparable output voltages, and significantly lower cost. For those applications where excellent linearity and stability are required, bulk-material Hall generators are recommended.

## Typical Applications Of Hall Generators

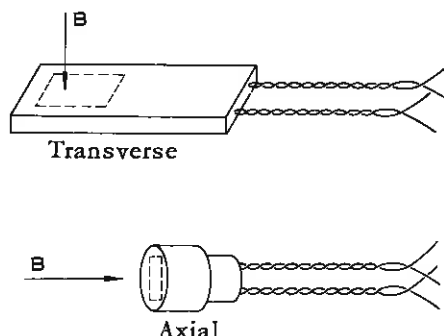
The following are just some of the many applications where Hall generators are used:

- Magnetic Card Readers
- Proximity Sensors
- Rotary Speed Sensors
- Watt Measurement
- Multipliers
- Magnet Field Measurements
- Electrical Power Measurements
- Current Sensors
- Brushless dc Motors
- Compasses
- Gaussmeters
- Watt-hour Meters
- Permanent Magnet Measurements
- Air Gap Measurements
- Magnetic Circuit design
- Flux Leakage Measurements

- Nondestructive Memory Readouts
- Linear/Angular Transducers
- Magnetic Tape Heads
- Guidance Systems
- Ignition Systems

### Typical Shapes and Sizes

Hall generators are available in a wide variety of shapes and sizes for adaptability to many different applications. The two basic types are transverse and axial, as illustrated in Figure 4.



**Figure 4**  
The two basic types of Hall generators are transverse and axial.

The transverse type is useful where the field must be measured in thin gaps and for multiplier applications. The axial

type must be used where the field is parallel to the axis of a hole, such as in traveling wave tubes or solenoids. Standard transverse probes as thin as .006" and axial probes as small as .063" in diameter are available.

Bulk-material Hall plates may be sandwiched between ferrite pieces to obtain effective air gaps less than .003". This may be useful in applications requiring maximum magnetic efficiency, such as electronic compasses and proximity sensors.

For a Hall generator to accurately measure flux density, the Hall plate area should be smaller than the cross section of the field to be measured. The output voltage is proportional to flux density, but a Hall plate is not equally sensitive over its entire area. If a high resolution is important, the Hall plate area should be small. Active areas as small 0.010" are available, while even smaller ones have been made. Units with somewhat larger Hall plates are usually less expensive because they are easier to make, and since they can generally handle larger currents they can produce more output voltage and dissipate more power.

4. Epstein, M., et al, "Principles and Applications of Hall-Effect Devices", Proceedings of the National Electronics Conference, 1959, Vol. 15, p. 241.  
5. Final Engineering Report on Hall Effect Device Investigation", Device Development Corporation, Weston 93, Massachusetts, Contract No. NOBsr-72823, July 1, 1958 to February 28, 1959, pp. 12-17.

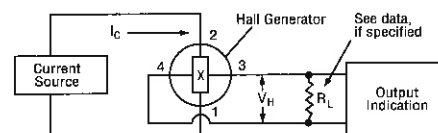
## ■ USING A F.W. BELL HALL GENERATOR

### CAUTION!

To avoid possible permanent damage to the Hall generator, please read the following instructions carefully before making connections to a power supply.

### Connecting The Hall Generator

The following schematic diagram illustrates the proper connections for the Hall generator:



**Figure 5**  
Hall generator Circuit Configuration.

**Lead** 1 and 2 are control current ( $I_C$ ) leads  
3 and 4 are Hall voltage ( $V_H$ ) leads  
**Color Code:** AWG 34 red ( $+I_C$ ), black ( $-I_C$ )  
blue ( $+V_H$ ), yellow ( $-V_H$ )  
AWG 36 neutral ( $+I_C$ ), green ( $-I_C$ ) red ( $+V_H$ ), neutral ( $-V_H$ )

Refer to the Hall generator specification for the AWG size of the leads. If a loading resistor,  $R_L$ , is specified, then it must be added to the output circuit as shown in Figure 5 to obtain the specified linearity.

### Current Source

A constant current supply is recommended for applications requiring fixed control current. This eliminates effects of input resistance changes resulting from temperature or field variations (magnetoresistance effect). A "brute-force" constant current source may be made by connecting a large resistor (30 times  $R_{in}$  or higher) in series with a battery or constant voltage power supply. In any case, the short-circuit current should be within the maximum current rating of the Hall generator. The control current may be either ac or dc. This is determined by the nature of the field and the type of output signal desired.

### Output Indicator

The Hall output voltage,  $V_H$ , may be observed on any suitable instrument such as a millivolt meter, oscilloscope, or recorder. The input impedance of the instrument should be greater than approximately 1,000 ohms.

## CAUTION! ISOLATION REQUIRED!

Since the four Hall generator leads connect to four points on a semiconductor plate having different potentials, no two leads can be connected together without upsetting the operation. Therefore, the current source and the output indicator cannot have a common connection, but must be isolated from each other. One or the other, but not both, may be grounded.

## Misalignment (Null) Voltage Compensation

In the manufacturing of the Hall generator, the Hall voltage contacts are placed on the semiconductor plate as accurately as possible so that very little output voltage will exist when there is no magnetic field present. For many applications, this resistive null voltage is low enough to be neglected, but for low field applications, it may be appreciable compared to the Hall output voltage  $V_H$ . If this is the case, a null voltage balancing network such as that in Figure 6 will make it possible to reduce the resistive null voltage to zero. The fine control may not be required.

## Affects of Residual Magnetism

Care should be taken to ensure that what appears to be an offset voltage of the Hall generator is not really the result of a residual magnetic field. Any magnetic material with a residual field in close proximity to the Hall generator could effect a slight Hall output voltage,  $V_H$ . Items such as fixtures, jigs, probes, metal tables, metal cabinets, etc., are potential sources of residual magnetic fields. Even the Earth's magnetic field (approximately 1/2 gauss) could cause an undesirable "offset" voltage. The circuit in Figure 6 can also be used to zero out many of these voltages.

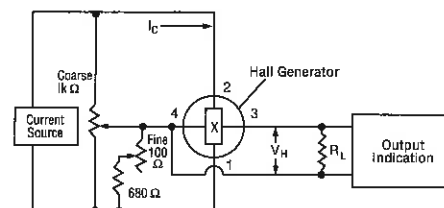


Figure 6  
Null Voltage Compensation

# HALL GENERATOR MOUNTING

## Handling

The Hall generator is fragile. It cannot be handled the same way most other electronic components are handled. The aluminum oxide substrate is brittle, thin and very sensitive to bending stress. Use the leads to move and locate it. Do not handle the substrate. The lead-to-substrate bond strength is on the order of an ounce. Avoid tension on the leads and avoid bending them close to the substrate. The leads may be bent at any angle as long as the bend is at least 1/8" away from the substrate connection.

## Slot Mounting

The preferred mounting procedure is to locate the chip in a slot that is any depth, .003 inch wider and .010 inch longer than the substrate. Tack the leads outside the slot with Sylgard 186\* or a similar substance. Don't get Sylgard 186 inside the slot. If an extreme temperature range is expected, check the coefficients of thermal expansion to be certain that the slot will always have clearance for the chip. This procedure is not recommended for installations that will be subject to any acceleration greater than 10 g.

## Surface Mounting

Surface mounting is acceptable when necessary. The mounting surface may be any non-flexible solid with a flat, smooth ( $\pm .001"$ ) surface at least the size of the substrate. The substrate must not overhang the mounting surface. Steel,

ferrite, ceramic, and glass are examples of mounting surfaces. For extended temperature ranges, choose a material with a coefficient of thermal expansion no greater than a factor of three different from that of the aluminum oxide substrate  $\approx 7 \times 10^{-6} \frac{IN}{IN \cdot ^\circ C}$

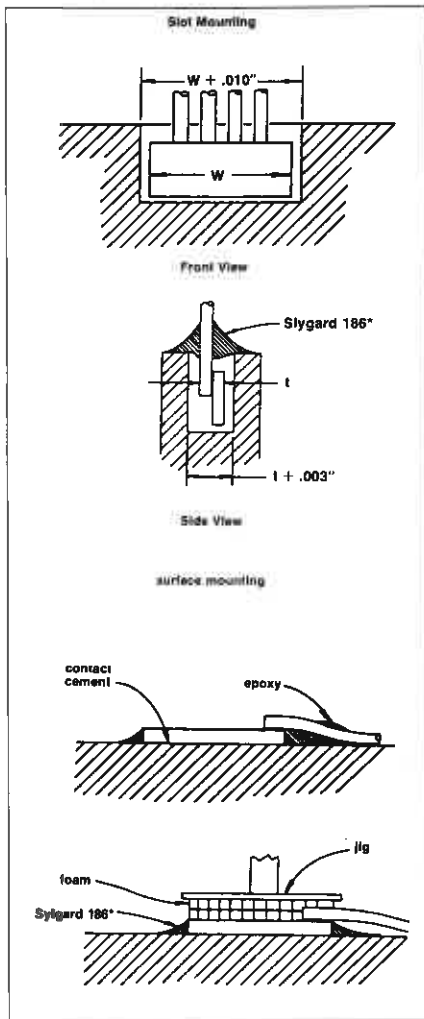
For a permanent mount, sparingly coat the mounting surface with Eastman 910 contact cement or other similar cement. The ceramic side of the substrate is visible as non-red or as opposite the Hall element. Locate the ceramic side on the clean, degreased surface and apply extremely light pressure with a foam pad until the bond is made (see Figure 7).

Wipe off the excess contact cement. Use an epoxy such as Bacon Industries FA8 or Emerson and Cuming 2850FT to form a fillet around the plate and to secure the leads. Don't get epoxy on top of the chip. If encapsulation is absolutely necessary, use a light coating of Sylgard 186 or a similar soft material.

For a non-permanent surface mount, secure the substrate against the surface with a foam-padded mounting jig. The jig should apply only light pressure. Temporarily secure the leads with Sylgard 186 or a similar material.

## Post Mounting Test

After the Hall generator has been mounted, check the misalignment voltage per the proper specification. A large misalignment voltage shift ( $100 \mu V$  or more) is a sign of Hall generator physical damage.

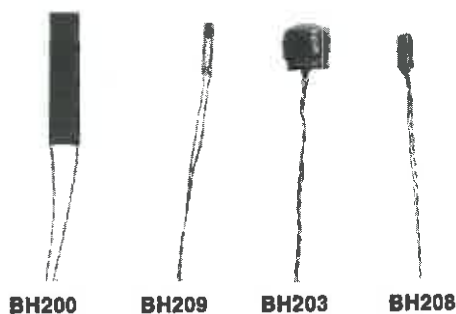


**Figure 7:**  
Mounting Configurations

\*Product of Dow Corning Corporation

# BH-200 SERIES

## INSTRUMENTATION QUALITY



(Above is only a portion of models available)

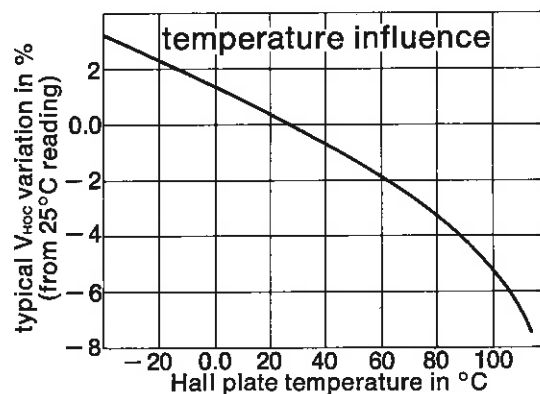
### General Description

The BH-200 series of Hall effect magnetic field sensors consists of ten models designed to meet the requirements of most magnetic field measurement applications.

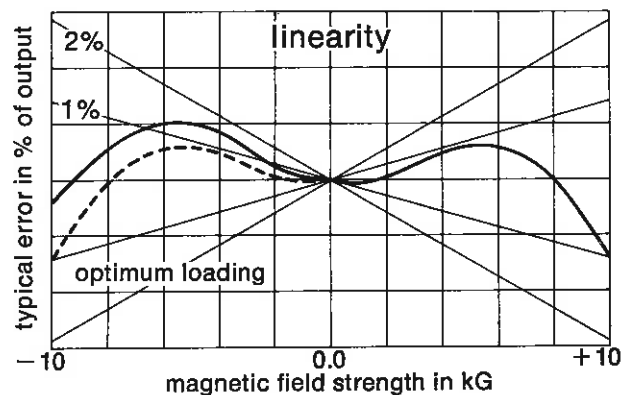
Models in the BH-200 Series are built in various configurations to measure axial, transverse, and tangential magnetic field components. Sensitivities range from 6 to 60 mV/kG with input and output resistance of several ohms.

### Models: Features:

BH-200	General Purpose Transverse
BH-201	Ultra-thin, Transverse
BH-202	Small Axial
BH-203	General Purpose, Axial
BH-204	Mini Axial
BH-205	Mini Transverse
BH-206	High Sensitivity, Low-cost Transverse
BH-207	High Resolution, Tangential
BH-208	Ultra-mini, Axial
BH-209	Ultra-mini, Transverse



NOTE: For an unmounted Hall device supported by it's leads, typical Hall plate temperature rise is 20° C for nominal control current.



NOTE: The dotted line is a mirror image of the curve in the right hand plane and illustrates the reversibility error.



# BH-200 SERIES

## INSTRUMENTATION QUALITY

### MECHANICAL SPECIFICATIONS

#### POLARITY

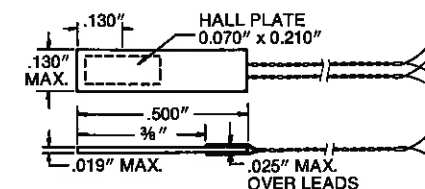
With the magnetic field vector (+B) entering the top of the Hall plate and  $I_C$  entering the red lead, the positive Hall voltage will appear at the blue lead.

#### LEADS

MATERIAL: AWG34 or AWG36 copper with heavy polyurethane insulation.  
 COLOR CODE: Control Current ( $I_C$ ): AWG34-red (+ $I_C$ ), black (- $I_C$ ), AWG36-neutral (+ $I_C$ ), green (- $I_C$ )  
 HALL VOLTAGE: ( $V_H$ ): AWG34-blue (+ $V_H$ ), yellow (- $V_H$ ), AWG36-red (+ $V_H$ ), neutral (- $V_H$ )

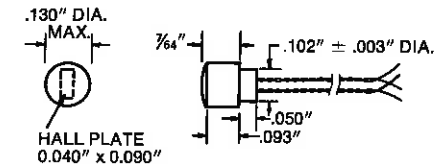
### ELECTRICAL SPECIFICATIONS

MODEL BH-200: GENERAL-PURPOSE TRANSVERSE



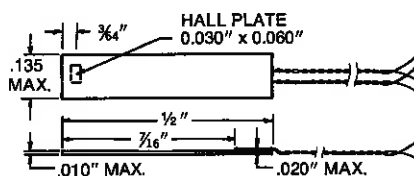
LEADS: AWG 34, 10" long

MODEL BH-202 SMALL AXIAL



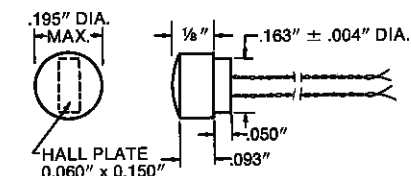
LEADS: AWG 36, 10" long

MODEL BH-201 ULTRA-THIN TRANSVERSE



LEADS: AWG 36, 10" long

MODEL BH-203 GENERAL-PURPOSE AXIAL



LEADS: AWG 34, 10" long

MODELS	BH-200	BH-201	BH-202
Rated control current, $I_{CN}$	150mA	100mA	100mA
Maximum continuous control current, $I_{Cmax}$ (in 25°C static air)	250mA	150mA	150mA
Open circuit magnetic sensitivity, $\gamma_B$ (at rated $I_{CN}$ , $B=1.0$ kG)	15mV/kG ±25%	12mV/kG ±25%	10mV/kG ±25%
Misalignment voltage, $V_M$ ( $B=0$ , $I_C=100$ mA)	100μV max	250μV max	100μV max
Inductive null constant, A (see note)	.003 cm <sup>2</sup> *	.01 cm <sup>2</sup> *	.002 cm <sup>2</sup> *
Hall output load resistance, $R_{lin}$ (for optimum linearity 0 to 10 kG)	15Ω*	15Ω*	15Ω*
Linearity error with $R_{lin}$ termination (0 to 10 kG)	1% max	1.5% max	1% max
Product sensitivity, $\gamma_B$ ( $B=10$ kG)	0.10V/A•kG ±25%	0.12V/A•kG ±25%	0.10V/A•kG ±25%
Input resistance, $R_{in}$ (including leads)	2.0Ω*	2.3Ω*	2.6Ω*
Output resistance, $R_{out}$ (including leads)	1.7Ω*	1.8Ω*	1.8Ω*
Mean temperature coefficient of $V_{HOC}$ (-10°C to +80°C)	-0.08%/°C*	-0.08%/°C*	-0.08%/°C*
Mean temperature coefficient of resistance (-10°C to +80°C)	0.15%/°C*	0.15%/°C*	0.15%/°C*
Reversibility error of $V_{HOC}$ (0 to 10 kG)	1% max	3% max	1% max
Mean temperature coefficient of misalignment voltage, ( $I_C=100$ mA), $D_T$	1μV/°C max	1μV/°C max	1μV/°C max
Storage temperature range	-40°C to +105°C	0.0 to +50°C	-40°C to +105°C
Operating temperature range (at rated $I_C$ )	-40°C to +100°C	0.0 to +50°C	-40°C to +100°C

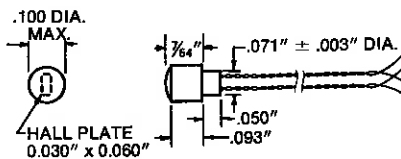
\*Approximate  
 \*\*Constant  $I_C$

NOTE: In a time varying field the voltage induced into the Hall output leads,  $V_{ind}$ , is proportional to the effective area, A, of the Hall output loop and the amplitude and the rate of change of the field,  $V_{ind}$  (measured with  $I_C=0$ ) =  $A \frac{dB}{dt} \times 10^{-8}$

$V_{ind}$  = volts, A = cm<sup>2</sup>, B = gauss, t = sec.

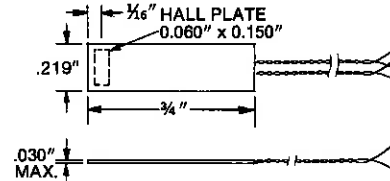


MODEL BH-204 MIDGET AXIAL



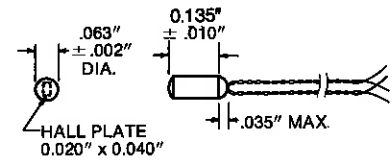
LEADS: AWG 36, 10" long

MODEL BH-206 HIGH SENSITIVITY LOW COST TRANSVERSE



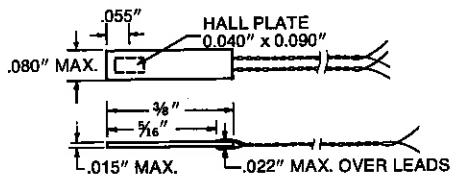
LEADS: AWG 34, 10" long

MODEL BH-208 ULTRA-MIDGET AXIAL



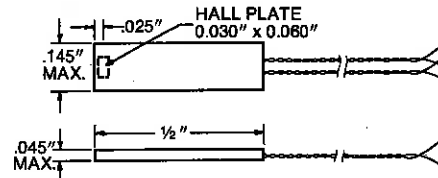
LEADS: AWG 36, 10" long

MODEL BH-205 MIDGET TRANSVERSE



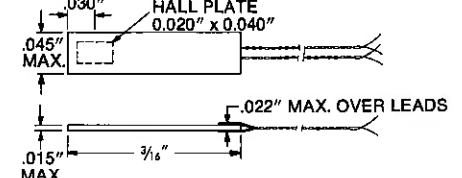
LEADS: AWG 36, 10" long

MODEL BH-207 HIGH RESOLUTION TANGENTIAL



LEADS: AWG 36, 10" long

MODEL BH-209 ULTRA-MIDGET TRANSVERSE



LEADS: AWG 36, 10" long

BH-203	BH-204	BH-205	BH-206	BH-207	BH-208	BH-209
100mA	100mA	125mA	200mA	150mA	100mA	75mA
250mA	150mA	200mA	250mA	250mA	150mA	150mA
10mV/kG ±25%	11mV/kG ±25%	12.5mV/kG ±25%	55mV/kG ±25%	16mV/kG ±25%	10mV/kG ±25%	6.75mV/kG ±25%
100μV max	200μV max	100μV max	500μV max	200μV max	250μV max	100μV max
.003 cm <sup>2</sup> *	.002 cm <sup>2</sup> *	.002 cm <sup>2</sup> *	.006 cm <sup>2</sup> *	.002 cm <sup>2</sup> *	.02 cm <sup>2</sup> *	.003 cm <sup>2</sup> *
15Ω*	15Ω*	15Ω*	25Ω*	15Ω*	15Ω*	15Ω*
1% max	1.5% max	1% max	2%*	1.5% max	1.5% max	1.5% max
0.10V/A·kG ±25%	0.11V/A·kG ±25%	0.10V/A·kG ±25%	.30V/A·kG ±25%	0.11V/A·kG ±25%	0.10V/A·kG ±25%	0.10V/A·kG ±25%
1.9Ω*	2.7Ω*	2.1Ω*	5Ω*	2Ω*	3.0Ω*	2.5Ω*
1.3Ω*	1.9Ω*	2.0Ω*	3.5Ω*	1.5Ω*	3.0Ω*	2.0Ω*
-0.08%/°C*	-0.08%/°C*	-0.08%/°C*	-0.2%/°C*	-0.08%/°C*	-0.08%/°C*	-0.05%/°C*
0.15%/°C*	0.15%/°C*	0.15%/°C*	0.2%/°C*	0.15%/°C*	0.15%/°C*	0.15%/°C*
1% max	1% max	1% max	1.5% max	1% max	1.5% max	1% max
1μV/°C max	1μV/°C max	1μV/°C max	6μV/°C max	1μV/°C max	1μV/°C max	0.5μV/°C max
-40°C to +105°C	-40°C to +105°C	-40°C to +105°C	-40°C to +105°C	-40°C to +105°C	-40°C to +105°C	-40°C to +105°C
-40°C to +100°C	-40°C to +100°C	-40°C to +100°C	-40°C to +100°C	-40°C to +100°C	-40°C to +100°C	-40°C to +100°C

### INAs THIN FILM, GENERAL PURPOSE, TRANSVERSE

FH-301 & FH-500 Series Hall generators are miniature solid-state Hall effect magnetic field sensing devices. The FH-500 series uses a lead strip which is composed of printed circuit leads encased in DuPont's Kapton and terminating in contacts on .075" centers. This flexible and tough lead strip provides for exceptionally easy handling. The lead strip can be made in a variety of configurations. The model FH-301 has conventional wire leads. The Models FH-301 and FH-500 are available in two different sensitivity ranges: 10.0 mV/kG and 12.0 mV/kG.

Technical drawings of three vertical rods, likely for a mechanical assembly. The drawings include the following dimensions and labels:

- Left Rod:**
  - Top flange width: 1.25"
  - Flange label: HALL PLATE 0.040" X 0.080"
  - Length: 7 1/2"
  - Bottom color coding: Vw (-) YELLOW, k (-) RED, Vw (+) BLUE
- Middle Rod:**
  - Top flange width: 0.020" MAX
  - Flange label: B ±
  - Break symbol in the middle of the rod.
- Right Rod:**
  - Top flange width: 0.135"
  - Flange label: HALL PLATE 0.040" X 0.080"
  - Top flange label: 0.025" MAX B ±
  - Length segments: 0.500", 2.000", 3.438"
  - Bottom flange width: 0.280"
  - Bottom flange label: T ± T ± T ±
  - Bottom dimensions: 0.075", 0.025", 0.125", 0.010"

Leads:  
#34 AWG  
Copper  
With  
Polyurethane  
Insulation

FH-301-020 • Low Current  
FH-301-040 • Lead Acid

FH-520	• Lowest Cost
FH-540	• Low Current
	• On-Lead Strip

**MODEL FH-301 SERIES      MODEL FH-500 SERIES**  
NOTE: All tolerances unless specified are  $\pm 0.010"$



**Model FH-500 Series  
with Flexstrip™ lead**

**POLARITY:** With field direction (B+) as shown and  $I_c$  entering the  $I_c$  (+) terminal, the positive Hall voltage will appear at the  $V_H$  (+) terminal.

**NOTE:** Unless otherwise specified, all specifications apply at nominal control current with T 25°C. Heat sinking can enhance performance in several respects. For information concerning operating on a heat sink, such as the pole face of a magnet, contact F.W.Bell, Inc.

Specifications may change without notice



# Model GH-601

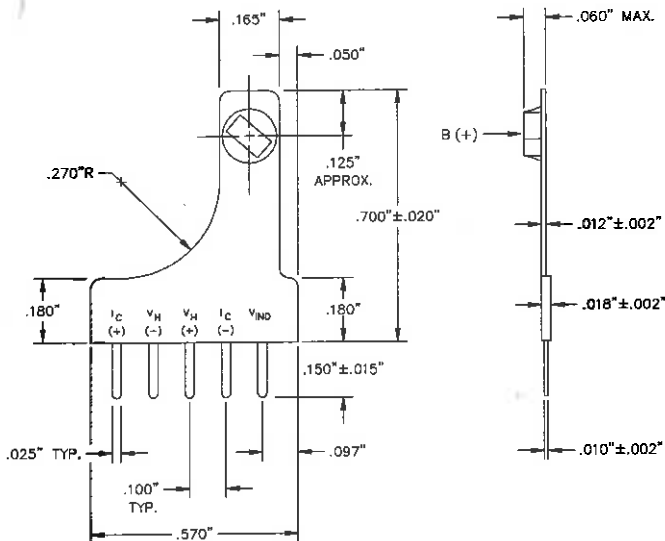
## Description:

The Model GH-601 Hall effect generator is a solid-state device that produces an output voltage,  $V_H$ , proportional to the product of the input current,  $I_C$ , and the magnetic flux density,  $B$ . A specially designed lead strip allows operation up to 50 kHz.

## Features:

- Low Cost
- Gallium Arsenide
- Extended Frequency Range
- High Sensitivity
- Low Current Requirement
- Flexible Lead Strip

**GH-601**  
MECHANICAL DIMENSIONS  
ALL DIMENSIONS ARE IN INCHES



Specifications	Units	GH-601
Input resistance, $R_{in}$	ohms	450 to 900
Output resistance, $R_{out}$	ohms	580 to 1700
Magnetic sensitivity, $V_H$ (1)	mV/kG	50 to 140
Max. resistive residual voltage, $V_M @ B = 0$ (1)	±mV	16
Maximum control current @ 25 °C, static air	mA	10
Nominal control current, $I_{cn}$	mA	5
Max. linearity error (-10 kG to +10 kG)	±% of RDG	2
Mean temperature coefficient of $V_H$ (-10 °C to +80 °C) (1)	%/°C	-0.07
Mean temperature coefficient of resistance (-10 °C to +80 °C) (1)	%/°C	0.15 Typical
Temperature dependence of resistive residual voltage (-10 °C to +80 °C) (2)	±μV/°C	1 Typical
Operating temperature range	°C	-55 °C to +125 °C
Storage temperature range	°C	-55 °C to +150 °C

## Notes:

- (1) Nominal Control Current,  $I_{cn}$  (5 mA)
- (2) Control current = 1 mA

As a result of continuous process improvement, specifications subject to change without notice



# Models GH-600/GH-700/GH-800

## Description:

The GH Series Hall effect generators are four-terminal solid-state devices that produce an output voltage,  $V_H$ , proportional to the product of the input current,  $I_C$ , and the magnetic flux density,  $B$ . The GH-600 Hall generator uses a lead strip which is composed of DuPont's Kapton. The lead strip is terminated with tin plated copper alloy contacts spaced 0.100 " (2.54 mm) on center. The GH-700 is an ion implanted planar device encased in an epoxy surface-mount package. The GH-800 is a leaded device designed for through hole mounting to a PCB. It features a package 0.28 " (0.7 mm) thick for placement in small air gaps.

## Features:

- Low Cost
- Gallium Arsenide
- Extended Temperature Range
- High Sensitivity
- Low Current Requirement
- Choice of Mounting Configurations

Specifications	Units	GH-600	GH-700	GH-800
Input resistance, $R_{in}$	ohms	450 to 900	450 to 900	600 to 1200
Output resistance, $R_{out}$	ohms	580 to 1700	580 to 1700	600 to 1200
Magnetic sensitivity, $V_H$ (1)	mV/kG	50 to 140	50 to 140	95 to 130
Max. resistive residual voltage, $V_M @ B = 0$ (1)	±mV	16	16	20
Maximum control current @ 25 °C, static air	mA	10	10	7
Nominal control current, $I_{CN}$	mA	5		
Max. linearity error (-10 kG to +10 kG)	±% of RDG	2	2	0.7 (4)
Mean temperature coefficient of $V_H$ (-10 °C to +80 °C) (1)	%/°C	0.07		
Mean temperature coefficient of resistance (-10 °C to +80 °C) (1)	%/°C	0.15 Typical	0.15 Typical	0.18 Max.
Temperature dependence of resistive residual voltage (-10 °C to +80 °C)	±µV/°C	1 Typical (2)	1 Typical (2)	40 Max. (1,3)
Operating temperature range	°C	-55 to +125	-55 to +125	-40 to +175
Storage temperature range	°C	-55 to +150	-55 to +150	-50 to +180

## Notes:

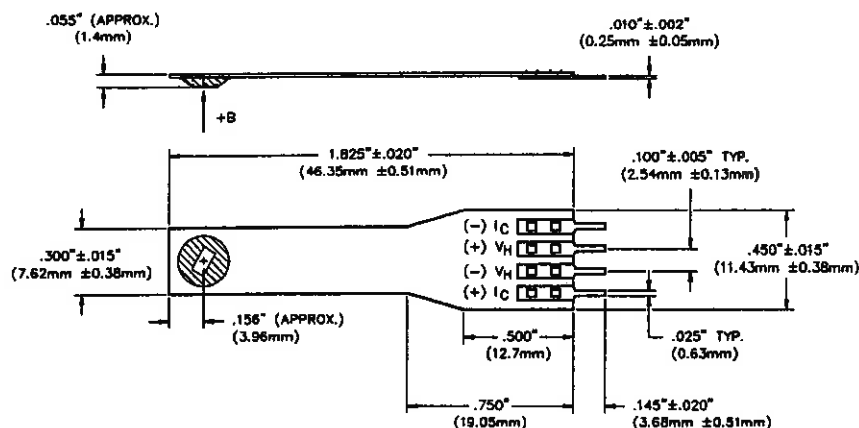
- (1) Nominal Control Current,  $I_{CN}$  (5 mA)
- (2) Control current = 1 mA
- (3) Temperature range +25 °C to +75 °C
- (4) ± 0.2% of reading from - 5 kG to + 5 kG

As a result of continuous process improvement, specifications subject to change without notice

# Models GH-600/GH-700/GH-800

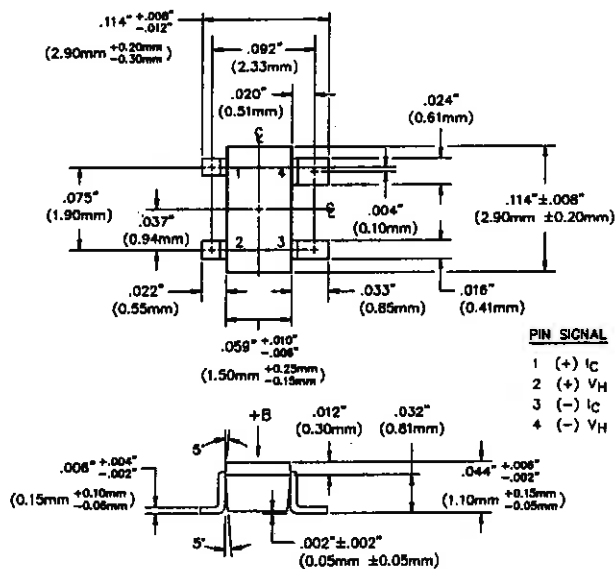
## MECHANICAL DIMENSIONS

ALL DIMENSIONS ARE IN INCHES (MILLIMETERS)

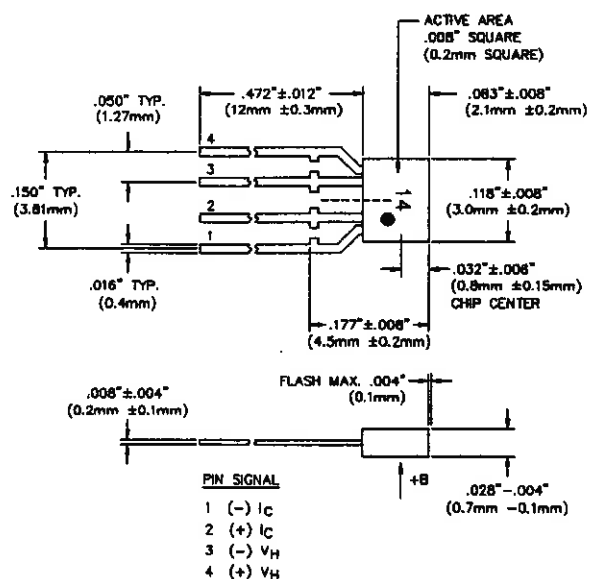


UNLESS OTHERWISE NOTED, ALL TOLERANCES ARE  $\pm 0.010"$  ( $\pm 0.25$ mm)

## MODEL GH-600



## MODEL GH-700



## MODEL GH-800

As a result of continuous process improvement, specifications subject to change without notice

# Models GH-810/GH-820/GH-830

## Description:

The GH Series Hall effect generators are four-terminal solid-state devices that produce an output voltage,  $V_H$ , proportional to the product of the input current,  $I_C$ , and the magnetic flux density,  $B$ . The GH-810 and GH-820 are leaded devices designed for through hole mounting to a PCB. The GH-830 is configured in a low profile package.

## Features:

- Low Cost
- Gallium Arsenide
- Extended Temperature Range
- High Sensitivity
- Low Current Requirement
- Choice of Mounting Configurations

Specifications	Units	GH-810	GH-820	GH-830
Input resistance, $R_{in}$	ohms	400 to 700	450 to 900	450 to 900
Output resistance, $R_{out}$	ohms	aprox. 2,000	3,200 max.	aprox. 3,000
Magnetic sensitivity, $V_H$ (1)	mV/kG	22 to 31	80 to 190	65 to 170
Max. resistive residual voltage, $V_M @ B = 0$ (1)	±mV	5	20	25
Maximum control current @ 25 °C, static air	mA	15	10	10
Nominal control current, $I_{cn}$	mA	5		
Max. linearity error (-10 kG to +10 kG)	±% of RDG	2		
Mean temperature coefficient of $V_H$ (-10 °C to +80 °C) (1)	%/°C	-0.05	-0.06	-0.05
Mean temperature coefficient of resistance (-10 °C to +80 °C) (1)	%/°C	0.5 Max. (2)	0.15 Typical	0.3 Max.
Temperature dependence of resistive residual voltage (-10 °C to +80 °C)	±µV/°C	1 Typical (2)	1 Typical (2)	5 Typical (2)
Operating temperature range	°C	-55 to +125		
Storage temperature range	°C	-55 to +150		

## Notes:

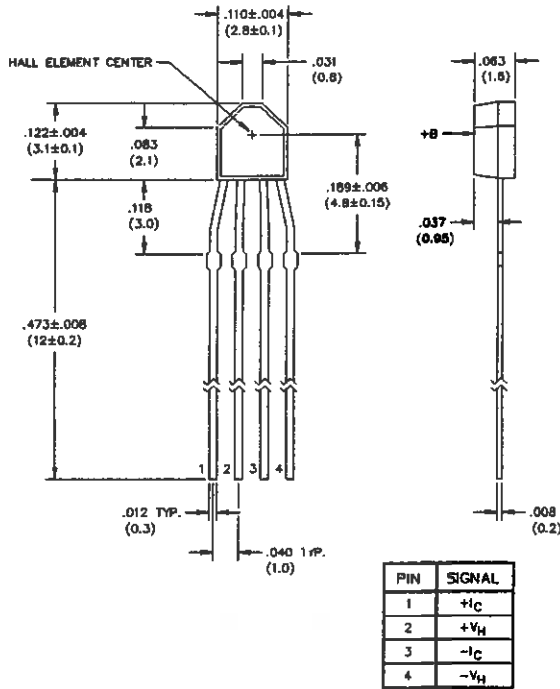
- (1) Nominal Control Current,  $I_{CN}$  (5 mA)  
 (2) Control current = 1 mA

As a result of continuous process improvement, specifications subject to change without notice

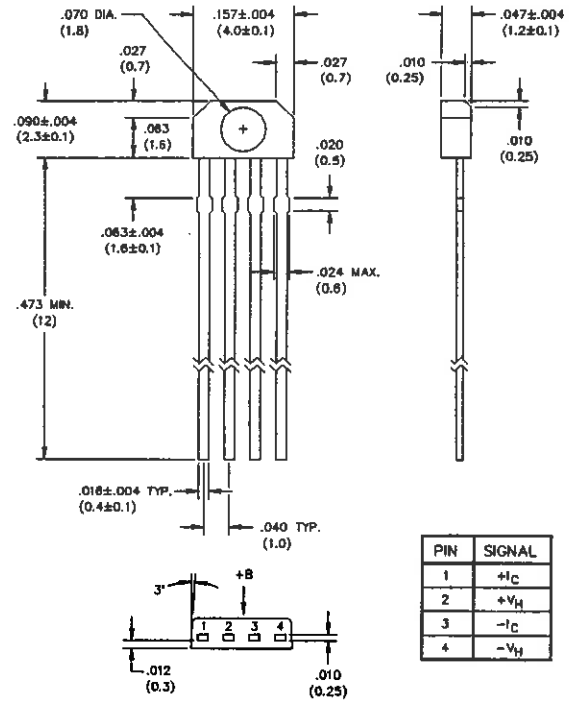
# Models GH-810/GH-820/GH-830

MECHANICAL DIMENSIONS  
ALL DIMENSIONS ARE IN INCHES (MILLIMETERS)

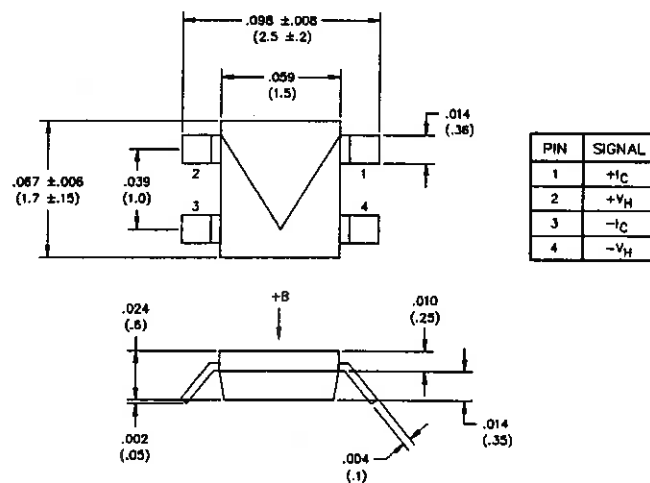
GH - 810



GH - 820



GH - 830

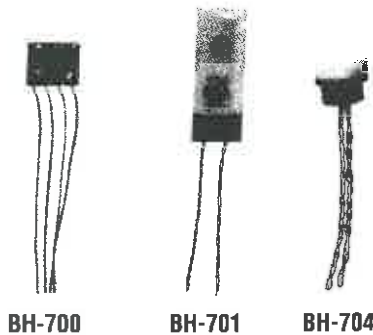


As a result of continuous process improvement, specifications subject to change without notice



# BH-700 SERIES

## SINGLE AXIS



Above is only a portion of the models available

### General Description:

Designed to meet the requirements of a wide range of magnetic field measurement applications, the BH-700 Series are small, solid-state devices that provide an output voltage proportional to the product of control current and ambient flux density.

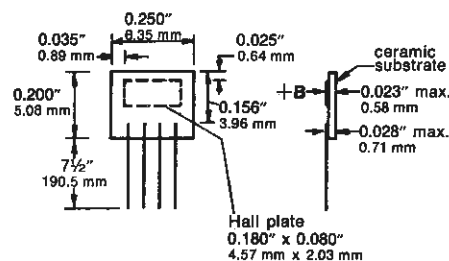
Five single-axis models are available to measure axial and transverse magnetic field components with sensitivities from 7.5 to 50 mV/kG and input and output resistance of several ohms.

### Models Features

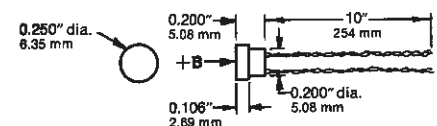
BH-700	Low Cost, Transverse, General Purpose
BH-701	Rugged, High-Linearity, Transverse, Instrumentation Quality
BH-702	Low Field (ferrite-embedded), Transverse
BH-704	Rugged, High Linearity, Axial, Instrumentation Quality
BH-705	General Purpose, Transverse

## MECHANICAL SPECIFICATIONS

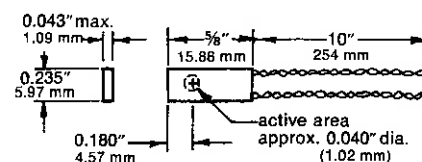
Model BH-700 Low cost transverse



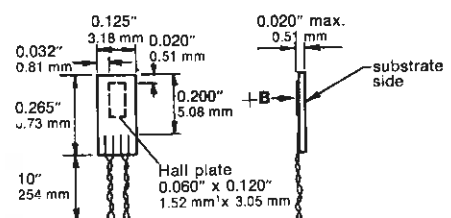
Model BH-704 high linearity axial



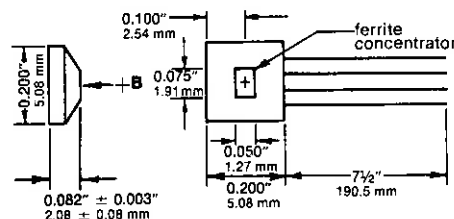
Model BH-701 High linearity transverse



Model BH-705 General purpose transverse



Model BH-702 ferrite imbedded transverse



NOTE: All tolerances unless specified are  $\pm 0.010"$   
Specifications may change without notice.

### COLOR CODE:

Control Current ( $I_C$ ): Red (+ $I_C$ )  
Black (- $I_C$ )  
Hall Voltage ( $V_H$ ): Blue (+ $V_H$ )  
Yellow (- $V_H$ )

### POLARITY:

With the magnetic field vector (+B) entering the top of the Hall plate and  $I_C$  entering the red lead, the positive Hall voltage will appear at the blue lead.

# BH-700 SERIES

## SINGLE AXIS

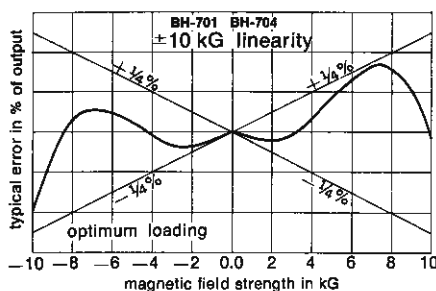
### ELECTRICAL SPECIFICATIONS

#### \*\*\*BH-702

**Air gap:** between concentrator and substrate, 0.0025" nominal and 0.003" maximum.

**Sensitivity:** With the unit suspended in a free field of 100 oersteds and 200 mA control current, the open circuit Hall voltage is 9.0 mV minimum. In a closed magnetic circuit driven with 2.5 ampere-turns,  $V_H$  is at least 8 mV with 200 mA control current.

**Polarity:** With the magnetic field vector as shown and  $I_C$  entering the red lead, the positive Hall voltage will appear at the blue lead.



NOTE: Optimum loading range for  $\pm 10$  kG operation is 20-50 $\Omega$ .

#### \*\*BH-701 & BH-704

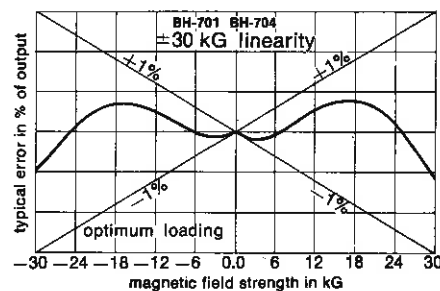
**Linearity:** (See below)

$V_H$  vs.  $B$ , -10 to +10 kG:  $\pm 0.25\%$  of reading, max.

$V_H$  vs.  $B$ , -30 to +30 kG:  $\pm 1.0\%$  of reading, max.

$V_H$  vs.  $I_C$ , 0 to 100 mA:  $\pm 0.1\%$  of reading, max.

$V_H$  vs.  $I_C$ , 0 to 300 mA:  $\pm 1.0\%$  of reading, max.



NOTE: Optimum loading range for  $\pm 30$  kG operation is 90-200 $\Omega$ .

**Encapsulation:** The BH-701 and the BH-704 are encapsulated in a rugged aluminum oxide ceramic and epoxy case for excellent heat transfer and strength.

JA0701

MODELS	BH-700	BH-701	BH-702	BH-704	BH-705
Rated control Current, $I_{CN}$	200 mA	100 MA	200 mA	100 mA	100 mA
Maximum continuous control current $I_{CMOS}(B \leq 3kG)$	250 mA	300 MA	300 mA	300 mA	250 mA
Product sensitivity	0.25 V/A•k min.	0.075 V/A•kG $\pm 20\%$	—	0.075V/A•kG $\pm 25\%$	0.1 V/A•kG $\pm 25\%$
Open circuit magnetic sensitivity, $\gamma_B$	50 mV/kG min.	7.5 mV/kG $\pm 20\%$ $I_C = I_{CN}$ 22 mV/kG $\pm 20\%$ $I_C = I_{CMOS}$	***	7.5 mV/kG $\pm 20\%$ $I_C = I_{CN}$ 22 mV/kG $\pm 20\%$ $I_C = I_{CMOS}$	10 mV/kG $\pm 25\%$
Misalignment voltage, $V_M(B = 0)$	1.5 mV max.	75 $\mu$ V max.	250 $\mu$ V max.	75 $\mu$ V max.	300 $\mu$ V max.
Linearity error with $R_{in}$ termination (% of reading)	3.0% 0 to 10 kG	**	—	**	1% 0 to 10 kG
Input resistance, $R_{in}$	5.5 $\Omega$ max.	2 $\Omega$ max.	3.5 $\Omega$ max.	2.5 $\Omega$ max.	2.2 $\Omega$ max.
Output resistance, $R_{out}$	5.5 $\Omega$ max.	2 $\Omega$ max.	3.5 $\Omega$ max.	2.5 $\Omega$ max.	2 $\Omega$ max.
Mean temperature coefficient of $V_{HOC}$ , $\beta_T$	-0.2%/°C* (-10°C to +80°C)	-0.04%/°C* (-10°C to +80°C)	-0.07%/°C* (-10°C to +60°C)	-0.04%/°C* (-10°C to +80°C)	-0.08%/°C max. (-10°C to +80°C)
Mean temperature coefficient of resistance	+0.20%/°C* (-10°C to +80°C)	+0.18%/°C* (-10°C to +80°C)	+0.18%/°C* (-10°C to 60°C)	+0.18%/°C* (-10°C to +80°C)	+0.2%/°C* (-10°C to +80°C)
Mean temperature coefficient of misalignment voltage	6 $\mu$ V/°C typ	0.3 $\mu$ V/°C typ	2.5 $\mu$ V/°C typ	0.5 $\mu$ V/°C max.	1 $\mu$ V/°C max.
Storage temperature range	-40°C to +105°C	-40°C to +105°C	-40°C to +105°C	-40°C to +105°C	-65°C to +105°C
Operating temperature range	-40°C to +100°C	-40°C to +100°C	-40°C to +75°C	-40°C to +100°C	-65°C to +100°C
Zero field thermal voltage	—	5 $\mu$ V max.	—	5 $\mu$ V max.	5 $\mu$ V max.
Thermal resistance, plate to ambient	—	0.1°C/mW*	—	—	0.5°C/mW*
Thermal resistance, plate to substrate	—	0.01°C/C/mW*	—	—	0.025°C/mW*

\*Approximate Specifications may change without notice. Unless otherwise noted:  $B = 1$  kG,  $I_C = I_{CN}$ ,  $T = 25^\circ\text{C}$ , Static air.



# BH-703

## THREE AXIS

### General Description:

The BH-703 multi-axis Hall generator consists of three individual Hall elements oriented in mutually perpendicular planes and encapsulated in a small epoxy package. This enables the BH-703 to produce voltages proportional to the three orthogonal components ( $B_x$ ,  $B_y$ ,  $B_z$ ) of a magnetic flux in any direction. Thus the BH-703 may be permanently mounted or arbitrarily oriented to sense fields in any direction.

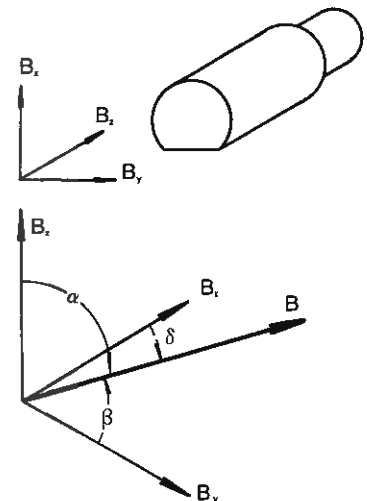
The magnitude of the flux vector,  $B$ , can be found using the following relation:  $B = \sqrt{B_x^2 + B_y^2 + B_z^2}$

The flux direction may be found using the following relations:

$\alpha = \cos^{-1} B_x/B$ ,  $\beta = \cos^{-1} B_y/B$ ,  $\delta = \cos^{-1} B_z/B$  where  $\alpha$ ,  $\beta$ ,  $\delta$  are the angles between  $B$  and  $B_x$ ,  $B_y$ ,  $B_z$  respectively.

### Model Features

- Three Axis, simultaneous measurement
- Instrumentation Quality

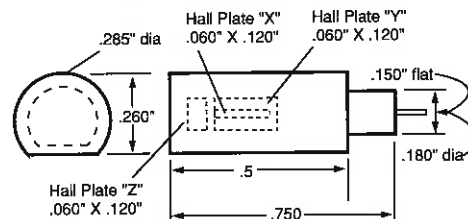


## MECHANICAL SPECIFICATIONS

### Multi-Axis Linearity

The linearity is within  $\pm 1.0\%$  of reading from  $-10$  kG to  $+10$  kG with resistive null voltage compensated to zero and with  $R_{IH}$  termination.

NOTE: All tolerances unless specified are  $\pm 0.010^\circ$ .



NOTE: Unless otherwise noted:  $B=1$  kG,  $I_C=I_{CN}$ ,  $T=25^\circ\text{C}$ , Static air.

Leads: #34 AWG copper with polyurethane insulation, approximately 20" long. The BH-703 has 12 leads.

Polarity: When the magnetic field vectors are oriented as shown, and  $I_C$  enters the red leads, the positive Hall voltage appears at the blue leads.

## ELECTRICAL SPECIFICATIONS

MODELS	BH-703
Zero field residual voltage $V_{MT}$ ( $B=0$ ), $I_C=100$ mA	100 $\mu\text{V}$ maximum
Angularity	Hall plates, perpendicular within $\pm 2^\circ$
Control current (a) nominal (b) max continuous	100 mA 300 mA
Input and output resistance, $B=0$	3 $\Omega$ maximum
Magnetic sensitivity (when terminated into optimum load, $R_{IH}$ )	7.5 mV/kG $\pm 20\%$
Sensitivity matching	within $\pm 1\%$
Temperature dependance (a) of Hall voltage (b) of resistance (c) zero field residual voltage	-0.04%/°C max. +0.15%/°C approx. 0.5 $\mu\text{V}/^\circ\text{C}$ max.
Operating temperature range	-40°C to +100°C

Specifications may change without notice.

# BH-706

## TWO AXIS

### General Description:

A single Hall element produces a voltage proportional to the magnetic flux density normal to its surface. The BH-706 multi-axis Hall generator consists of two Hall elements mounted in mutually perpendicular planes and encapsulated in a small epoxy package. This enables the BH-706 to produce voltages proportional to two perpendicular components ( $B_x$ ,  $B_y$ ) of a magnetic field. Thus the BH-706 may be permanently mounted to sense field components in its X, Y planes.

The magnitude of the flux vector,  $B$ ,

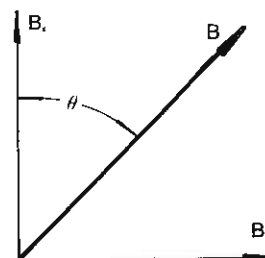
within the X,Y plane can be found using the following equation.

$$B = \sqrt{B_x^2 + B_y^2}$$

The direction of  $B$  can be computed using the following equation:

$$\theta = \tan^{-1} B_y / B_x$$

where  $\theta$  is the angle between  $B$  and  $B_x$ .



## MECHANICAL SPECIFICATIONS

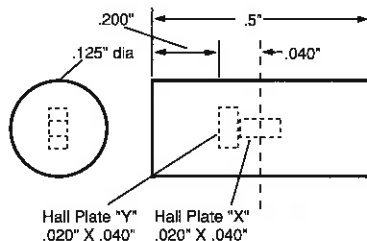
### Multi-Axis Linearity

The linearity is within  $\pm 1.0\%$  of reading from  $-10$  kG to  $+10$  kG with resistive null voltage compensated to zero and with  $R_{lin}$  termination.

Leads: #34 AWG copper with polyurethane insulation, approximately 20" long. The BH-706 has 8 leads.

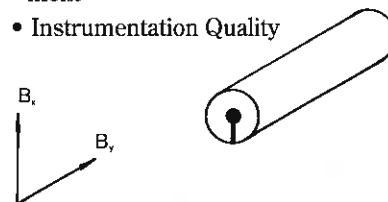
Polarity: When the magnetic field vectors are oriented as shown, and  $I_c$  enters the red leads, the positive Hall voltage appears at the blue leads.

**NOTE:** All tolerances unless specified are  $\pm 0.010"$ .



### Model Features:

- 2 Axis, simultaneous measurement
- Instrumentation Quality



Unless otherwise noted:  $B=1$  kG,  $I_c = I_{cN}$ ,  $T=25^\circ\text{C}$ , Static air.

## ELECTRICAL SPECIFICATIONS

MODELS	BH-706
Zero field residual voltage $V_{MT}$ ( $B=0$ ), $I_c=100$ mA	200 $\mu\text{V}$ maximum
Angularity	Hall plates, perpendicular within $\pm 2^\circ$
Control current (a) nominal (b) max continuous	100 mA 300 mA
Input and output resistance, $B=0$	35 $\Omega$ maximum
Magnetic sensitivity (when terminated into optimum load, $R_{lin}$ )	7.5 mV/kG $\pm 20\%$
Sensitivity matching	within $\pm 1\%$
Temperature dependence (a) of Hall voltage (b) of resistance (c) zero field residual voltage	-0.04%/°C max. +0.15%/°C approx. 0.5 $\mu\text{V}/^\circ\text{C}$ max.
Operating temperature range	-40°C to +100°C

Specifications may change without notice.



# BH-850

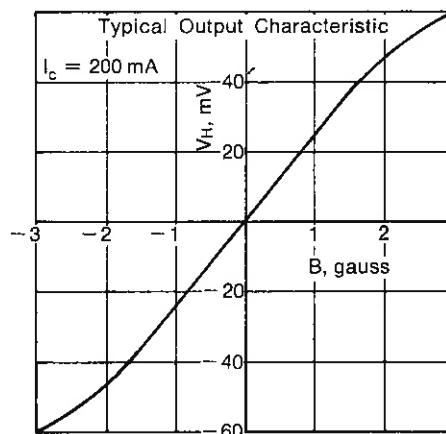
## ULTRA LOW FIELD

### General Description

Ideal for applications such as the construction of solid state compasses, the BH-850 offers high sensitivity for very low magnetic fields at a relatively low cost.

### Model Features

- High Sensitivity
- Rugged construction



### LEADS:

Material: AWG 34 Stranded (7x42) Silver Plated  
Copper, Thin Wall Teflon Insulation.

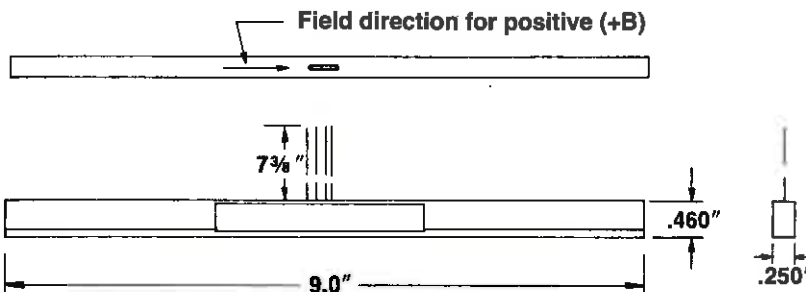
### COLOR CODE:

Control Current ( $I_c$ ): Red (+ $I_c$ )  
Black (- $I_c$ )  
Hall Voltage ( $V_H$ ): Blue (+ $V_H$ )  
Yellow (- $V_H$ )

### Polarity:

With field direction as shown  $I_c$  entering the red lead, the positive Hall output voltage will appear at the blue lead.

## MECHANICAL SPECIFICATIONS



## ELECTRICAL SPECIFICATIONS

MODEL	BH-850
Nominal control current, $I_{cn}$	200 mA
Maximum continuous control current, $I_{cmos}$ (in static air with $T=25^\circ\text{C}$ )	300 mA
Input resistance, $R_{in}$	$3.5\Omega$ max.
Output resistance, $R_{out}$	$3.5\Omega$ max.
Sensitivity, $B$ ( $I_c=200 \text{ mA}$ , open circuit)	18 mV/g. min.
Misalignment voltage, $V_M$ ( $\beta=0$ , $I_c=200 \text{ mA}$ , $T=25^\circ\text{C}$ )	$\pm 200 \mu\text{V}/^\circ\text{C}$ max.
Temperature dependence of misalignment voltage	$\pm 2.5 \mu\text{V}/^\circ\text{C}$ max.
Temperature coefficient of open circuit hall voltage, $V_{HOC}$ , from $-55^\circ\text{C}$ to $+85^\circ\text{C}$	$-0.18\%/^\circ\text{C}$ max.
Operating temperature range	$-55^\circ\text{C}$ to $+85^\circ\text{C}$
Storage temperature range	$-55^\circ\text{C}$ to $+85^\circ\text{C}$

Specifications may change without notice

# BH-900 SERIES

## HIGH LINEARITY

### LEADS:

Material: AWG 34 Copper with Teflon Insulation (Model 921) or Polyurethane Insulation (Models 900 & 910).

**COLOR CODE:** Control Current  $I_C$

Control Current ( $I_C$ ) Red (+ $I_C$ )  
Black (- $I_C$ )

Hall Voltage ( $V_H$ ) Blue (+ $V_H$ )  
Yellow (- $V_H$ )

### POLARITY

With the magnetic field vector (+B) entering the top of the Hall plate and  $I_C$  entering the red lead, the positive Hall voltage will appear at the blue lead.

### General Description:

F.W.Bell 900 Series Hall Generators are high-performance units providing high linearity and broad field and temperature ranges for a wide variety of magnetic field measurements. All units in the series are encapsulated in rugged, epoxy-sealed cases. A room temperature linearity error curve from -30 to +30

kG is supplied, indicating optimum operating conditions for each device. The models 900 and 921 are not calibrated above 30 kG.

### Models

BH-910

BH-921

BH 921 & 900

### Features:

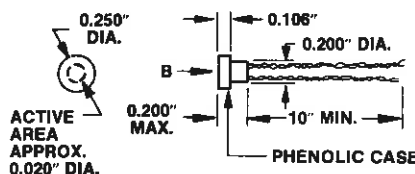
High Linearity

Cryogenic Operation  
(1.5 to 350° K)

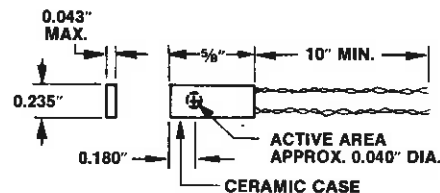
Wide Dynamic Range

## MECHANICAL SPECIFICATIONS

NOTE: Cross indicates tail of magnetic field vector.



Axial Hall Generators BHA-900, 910 & 921



Transverse Hall Generators BHT-900, 910 & 921

## ELECTRICAL SPECIFICATIONS

Note: Unless otherwise stated, all specifications apply at nominal control current with  $T=25^\circ\text{C}$ .

MODELS	BHT-900	BHT-910	BHT-921	BHA-900	BHA-910	BHA-921
Nominal control current, $I_{CN}$	100mA	100mA	100mA	100mA	100mA	100mA
Maximum continuous control current, $I_{CMOS}$ (in $25^\circ\text{C}$ static air)	300mA	300mA	300mA	300mA	300mA	300mA
Magnetic sensitivity, $\gamma_B$ , $I_C = 100$ mA	0.8mV/kG $\pm 30\%$	0.8mV/kG $\pm 30\%$	0.8mV/kG $\pm 30\%$	0.8mV/kG $\pm 30\%$	0.8mV/kG $\pm 30\%$	0.8mV/kG $\pm 30\%$
Typical load required for proper linearity	500 $\Omega$	50 to 500 $\Omega$	500 $\Omega$	500 $\Omega$	50 to 500 $\Omega$	500 $\Omega$
Linearity error ( $I_C = 100$ mA) -30 to +30kG	$\pm 1.0\%$ (max.)	$\pm 0.1\%$ (max.)	$\pm 1.0\%$ (max.)	$\pm 1.0\%$ (max.)	$\pm 0.25\%$ (max.)	$\pm 1.0\%$ (max.)
Linearity error ( $I_C = 100$ mA) -150 to +150kG	$\pm 1.5\%$ (max.)	—	$\pm 2.0\%$ (max.)	$\pm 1.5\%$ (max.)	—	$\pm 2.0\%$ (max.)
Operating temperature range	-40 to +100°C	-40 to 100°C	-269 to +100°C	-40 to +100°C	-40 to +100°C	-269 to +100°C
Mean temperature coefficient of Hall voltage, $\beta_T$	$\pm 50$ ppm/°C*	$\pm 50$ ppm/°C*	$\pm 100$ ppm/°C*	$\pm 50$ ppm/°C*	$\pm 50$ ppm/°C*	$\pm 100$ ppm/°C*
Mean temperature coefficient of resistive residual voltage, $D_T$	$\pm 0.4$ $\mu\text{V}/^\circ\text{C}$	$\pm 0.4$ $\mu\text{V}/^\circ\text{C}$	$\pm 0.4$ $\mu\text{V}/^\circ\text{C}$	$\pm 0.4$ $\mu\text{V}/^\circ\text{C}$	$\pm 0.4$ $\mu\text{V}/^\circ\text{C}$	$\pm 0.4$ $\mu\text{V}/^\circ\text{C}$
Mean temperature coefficient of resistance $\alpha_T$	$\pm 0.15\%/^\circ\text{C}$	$\pm 0.15\%/^\circ\text{C}$	$\pm 0.6\%/^\circ\text{C}$	$\pm 0.15\%/^\circ\text{C}$	$\pm 0.15\%/^\circ\text{C}$	$\pm 0.6\%/^\circ\text{C}$
Resistive residual voltage, $V_m$ , ( $I_C = 100$ mA)	50 $\mu\text{V}$ (max.)	50 $\mu\text{V}$ (max.)	200 $\mu\text{V}$ (max.)	50 $\mu\text{V}$ (max.)	50 $\mu\text{V}$ (max.)	200 $\mu\text{V}$ (max.)
Input resistance in zero field, $R_{in}$ (including leads)	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *
Output resistance in zero field, $R_{out}$ (including leads)	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *	1.0 $\Omega$ *

\* Approximate value. Specifications may change without notice.

# APPLICATION AIDS

## General Purpose Amplifier & Current Supply Circuit

Below is a schematic for a modest, linear, fixed-gain amplifier and simple current excitation scheme. This would be considered a moderate performance circuit. This circuit can be used with F.W.Bell Hall Generators. Refer to Table 1 for specific models.

**\*\*R<sub>1</sub> & R<sub>2</sub>:** R is adjusted to calibrate sensor to a fixed output sensitivity

**R<sub>4</sub> & R<sub>1</sub>:** Feedback resistors selected for desired gain

**R<sub>4</sub> & R<sub>5</sub>:** Input resistors should be 10k $\Omega$  (min.) to avoid loading the output

**R<sub>4</sub> & R<sub>1</sub>:** For proper common mode rejection (CMR)

Amplifier Frequency Response: dc to 10 kHz

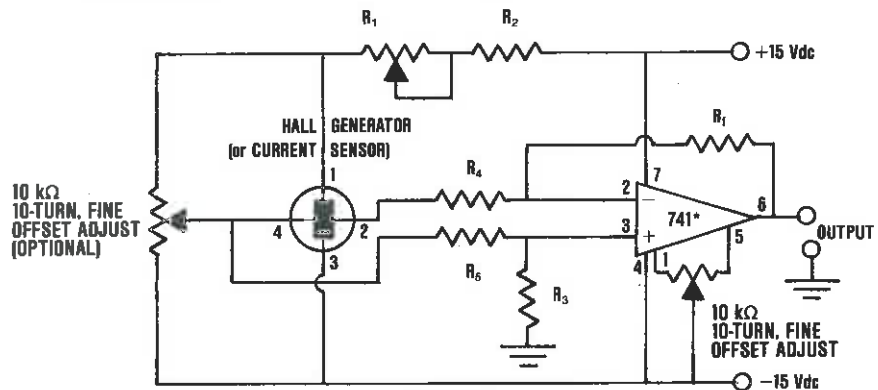


FIGURE 1

Note: Adjust offset pots, only when the Hall generator is in a zero gauss field

*\*The 741 operational amplifier is merely for the purpose of illustration, any other appropriate operational amplifier can be used.*

*\*\*For a true constant source that affords improved line regulation & temperature compensation, refer to "Constant Current Supply" below.*

TABLE 1 HALL GENERATOR CONNECTIONS

PIN OUT PER FIG. 1	1	2	3	4
BH-200, FH-300, BH-700, BH-900 SERIES	Red	Yellow	Black	Blue
FH-500 & GH-600 SERIES	+I <sub>c</sub>	-V <sub>H</sub>	-I <sub>c</sub>	+V <sub>H</sub>

## Constant current supply

On the following page is a typical schematic for a simple, low cost, constant current source that can be used to supply the excitation current needed to drive most thin film Hall generators. As shown, it has a variable current range of about 15 to 40mA dc. This variable supply can be adjusted to fix the output

sensitivity at a given current. This circuit has two inherent advantages:

- 1) It reduces the effects of sensitivity changes due to changes in ambient temperature to  $\leq 0.05\%/^{\circ}\text{C}$  ( $-40^{\circ}\text{C}$  to  $+80^{\circ}\text{C}$ ) maximum and 2) 1% variations of the input voltage (at +15V dc) will affect the accuracy by only 0.1%.

The circuits described herein are intended only to illustrate typical applications. This information is believed reliable, however, no responsibility is assumed for inaccuracies.

# APPLICATION AIDS

Constant Current Supply (continued)

For variations to this circuit to accommodate other Hall generators (for example: different input voltages, different current swings, different

regulator voltages, etc.), consult the data books and application notes furnished by the manufacturer of the voltage regulator.

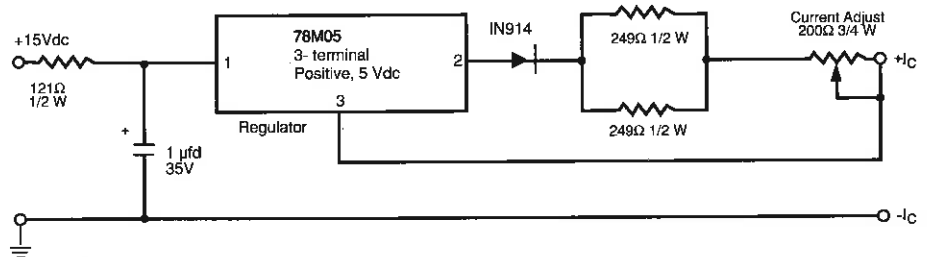


Figure 2

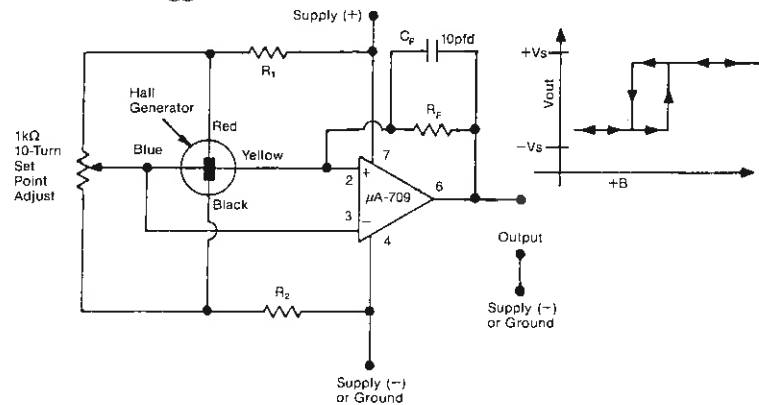
## Hall generator switching circuit

Below is a schematic for an easy to build switching circuit for use with an F.W.Bell, Inc. Hall generator. The circuit is triggered by the output voltage of the Hall generator, and the point at which the circuit is triggered is variable.

The amplitude of the circuit's output voltage is approximately:

$$V_{OUT} = 2 V_S - 2(V).$$

$R_1 = R_2$ : Selected for desired Hall generator control current  
 $R_F$ : Feedback resistor selected for the required hysteresis. Typical value is 100Ω  
 Frequency response: dc to 10 kHz



\* The 709 operational amplifier is merely for the purpose of illustration. Any other appropriate operational amplifier can be used.

Figure 3

\*Values for testing BH-Series Hall generators

\*\*Values for testing FH-Series Hall generators

\*\*\*Values for testing GH-Series Hall generators

The circuits described herein are intended only to illustrate typical applications. This information is believed reliable, however, no responsibility is assumed for inaccuracies.

## Testing F.W. Bell Hall generators

### 1. Test Current Calibration

(Refer to Fig. 4)

With switch S-1 in the "CAL" position, adjust the constant current supply output until the dc voltmeter

reads 100 mV\* (or 10mV\*\* or 1 mV\*\*\*). The test current is now calibrated at 100 mA\* (or 10 mA\*\* or 1 mA\*\*\*).



# APPLICATION AIDS

Testing F.W.Bell Hall generators (continued)

## 2. Input Resistance Measurement

Place the test Hall generator leads in the proper Kelvin Klips. Move switch S-1 to the "R<sub>in</sub>" position. The reading on the voltmeter in millivolts, divided by 100\* (or 10\*\* or 1\*\*\*), is the value of the input resistance.

## 3. Output Resistance Measurement

Same procedure as input resistance except readings taken with S-1 in position 3.

## 4. Zero Field Residual Voltage

With the hall generator still in the Kelvin Klips, move the selector switch to the "V<sub>H</sub>" position. Place the Hall generator active area into F.W.Bell Model YA-111 zero gauss chamber. The zero field residual voltage at a control current of 100 mA\* (or 10 mA\*\* or 1 mA\*\*\*) is read on the dc voltmeter.

The value at higher control current can be calculated from the fact that the zero field residual voltage is directly pro-

portional to the control current. If desired, the control current may be recalibrated, as in Step 1, to the desired value prior to reading zero field residual voltage.

## 5. Magnetic Sensitivity Measurement

Remember or record the polarity and value of the zero field residual voltage.

Place the Hall generator in a known magnetic field (1 kilogauss is convenient) with the red surface toward the north pole of the magnet. Subtract algebraically the zero field residual voltage from the field-generated reading. This gives the output voltage swing. Dividing the voltage swing by the field value gives the magnetic sensitivity of the Hall generator at the calibrated test current.

When tests are complete move S-1 to the "CAL" position, then remove the Hall generator.

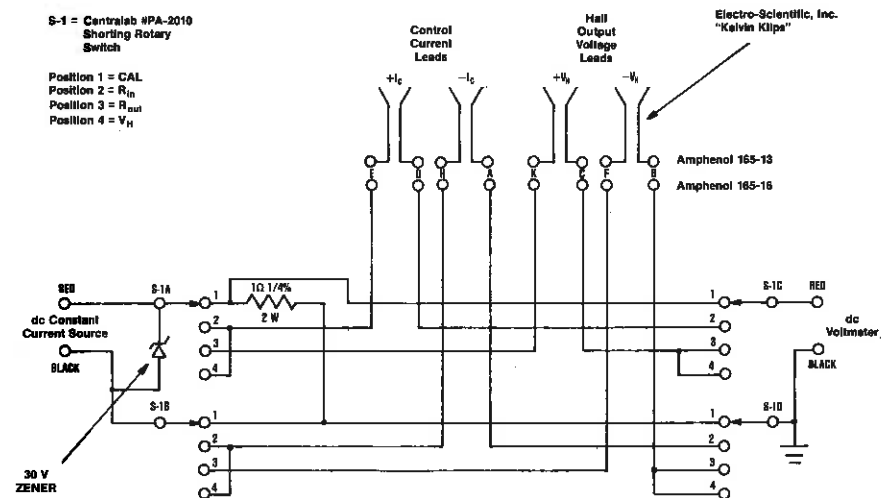


Figure 4

\*Values for testing BH-Series Hall generators  
 \*\*Values for testing FH-Series Hall generators  
 \*\*\*Values for testing GH-Series Hall generators

The circuits described herein are intended only to illustrate typical applications. This information is believed reliable, however, no responsibility is assumed for inaccuracies.

# APPLICATION AIDS

## Techniques for temperature compensating Hall generators

Temperature effects can rarely be ignored in the application of a Hall generator. Selecting a generator with a low-temperature coefficient of sensitivity may suffice, but compensation is occasionally required. If the designer understands the parameters involved, their effects on the Hall device output, and knows the ambient operating conditions, applying the compensating networks is simple. Due to the wide range of parameter variations among Hall generators of the same type, however, it is difficult to have one compensation network work for all generators that are used.

The compensation circuits offered below represent the basics; the total possibilities are limited only by the user's imagination. Only N-type indium arsenide Hall generators will be considered. Conduction is by majority carriers, no semiconductor junctions exist, and all contacts to the semiconductor material are ohmic.

An indium arsenide Hall generator usually exhibits a positive temperature coefficient of resistance ( $\alpha_T$ ) over its normal operating range. If a constant voltage supply provides the generator control current, this current decreases as the Hall plate temperature increases. The result is an output sensitivity inversely related to temperature. A constant current supply eliminates the effects of Hall generator input resistance changes. This condition can also be approximated by a voltage supply with high output impedance. If conditions

require a low-impedance voltage supply, the designer who wants to compensate for  $\alpha_T$  must be aware of a sign reversal on it, as shown in Figure 5.

For most hall generators with product sensitivities below 0.2 volts per ampere kilogauss, the sign change on  $\alpha_T$  takes place at temperatures above the normal operating range. In this case, the compensating networks discussed in the following sections may be used. Higher sensitivity devices exhibit the sign reversal within their operation range, however (see Figure 5). This behavior can make compensation very difficult, and the designer may have to accept the input resistance temperature effect or change to a higher source impedance.

## Sensitivity Variation

The relation of electron mobility to temperature causes Hall generator sensitivity to decrease as temperature increases. This phenomenon is independent of the temperature coefficient of resistance effect. Normally, Hall generators with higher sensitivities exhibit higher temperature coefficients of sensitivity ( $\beta_T$ ) at constant current. The output variation is approximately linear from  $-60^\circ$  to  $60^\circ\text{C}$ . Above  $60^\circ\text{C}$ , the rate of change increases rapidly (see Figure 6).

Several methods exist for compensating for temperature variation of the parameters discussed above. Since all the suggested procedures require trial-and-error calculations and testing, only the overall configuration is described.

A positive temperature coefficient (PTC) resistor can be used as a compen-

The circuits described herein are intended only to illustrate typical applications. This information is believed reliable, however, no responsibility is assumed for inaccuracies.

# APPLICATION AIDS

## Sensitivity Variation (continued)

sating element. (Figure 7). Nickel or copper wire resistance are possibilities, or PTC silicon resistors may be considered. Nickel exhibits a positive temperature coefficient of about  $0.5\%/^{\circ}\text{C}$ ; copper,  $0.4\%/^{\circ}\text{C}$ ; and silicon resistors, about  $0.7\%/^{\circ}\text{C}$ .

Negative temperature coefficient (NTC) thermistors are the most widely used devices for compensating Hall gen-

circuits usually exhibit good high-temperature correction. Improved compensation over a wider temperature range may be achieved through use of two temperature sensor networks. The two may be a combination of those mentioned above.

The following example (Figure 11) illustrates the results of negative temperature coefficient thermistor compensation

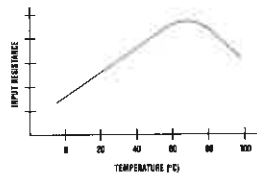


Figure 5. "Worst case" temperature characteristics of input resistance.

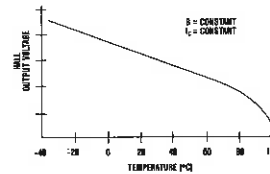


Figure 6. The sensitivity of a Hall generator decreases with the temperature

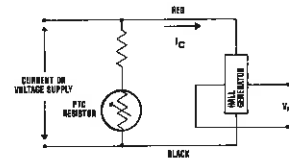


Figure 7. A positive temperature coefficient thermistor can be used in temperature compensation of Hall generators

Test results on sensitivity with compensated and uncompensated circuits.

erator temperature drift (Fig. 9). This configuration with a voltage supply causes the control current  $I_C$  to increase as the temperature rises.

Another compensation method, using an NTC thermistor, is shown in Figure 10. In implementing this concept, the designer must make sure that two conditions exist. Ample output must be maintained across resistor  $R_1$ , and the total load,  $R_L$ , presented to the Hall generator must adversely affect output linearity. (Consult the Hall generator manufacturer if linearity is a critical parameter.) If amplification is required, an alternative to the circuit in Figure 10 would be to place a thermistor circuit in the feedback network of the amplifier.

Single thermistor or PTC resistor

in the Hall generator input circuit. The Hall generator has the characteristics listed below:

Input Resistance ( $R_{in}$ ),  $4\Omega$   
Magnetic Sensitivity ( $\gamma_B$ ),  $30 \text{ mV/kG}$   
(At  $I_C = 100 \text{ mA}$  and  $T = 25^{\circ}\text{C}$ )  
Temperature Coefficient of Input Resistance ( $\alpha_T$ ),  $+0.2\%/^{\circ}\text{C}$   
Temperature Coefficient of Magnetic Sensitivity ( $\beta_T$ ),  $-0.22\%/^{\circ}\text{C}$

Improved Hall generator temperature drift is desired between  $25^{\circ}$  and  $60^{\circ}\text{C}$ . Uncompensated, the generator sensitivity at  $60^{\circ}\text{C}$  would decrease approximately 8% from its value at  $25^{\circ}\text{C}$ . A first attempt to improve this temperature drift might be the circuit shown in Figure 11. The real problem is to select proper values for  $R_1$  and  $R_T$ .

Increasing the Hall generator control

The circuits described herein are intended only to illustrate typical applications. This information is believed reliable, however, no responsibility is assumed for inaccuracies.

# APPLICATION AIDS

## Sensitivity Variation (continued)

current by about 8% at 60°C will eliminate the drop in sensitivity due to  $\beta_T$ . The next step is to calculate a thermistor-resistor parallel resistance,  $R_{TOT}$  which will supply 108 mA control current at 60°C. (Changes in resistor  $R_1$  and in the Hall generator input resistance are small and will be ignored throughout this example).

$$R_{TOT} = (V_S/I) - R_{IN} \\ = 5/(0.108) - 4 \approx 42.3\Omega$$

Since thermistor resistance will be selected at 60°C,  $R_T$  will generally be high

in value at 25°C. Assuming that  $R_1$  will be moderately high, let  $R_1$  be 47 $\Omega$ . If our assumptions are correct, the Hall generator control current will be near 100 mA at 25°C. Next, the thermistor resistance at 60°C must be calculated;

$$R_T = \frac{R_1(R_{TOT})}{R_1 - R_{TOT}}$$

The table gives a comparison of sensitivity change for both the uncompensated and compensated circuits (a thermistor temperature coefficient of -3.9%/°C was used). A higher  $R_1$  might be selected for better low-temperature response.

T (°C)	$\gamma$ (mV/kG)	$R_T$ ( $\Omega$ )	$R_{TOT}$ ( $\Omega$ )	I (mA)	$\gamma_B$ (comp.) (mV/kG)
-10	32.3	6447	46.66	98.7	31.88
+10	31.0	2636	46.18	99.6	30.88
+25	30.0	1440	45.51	101.0	30.30
+40	29.0	827	44.47	103.2	29.93
+60	27.7	423	42.30	108.0	29.92

Figure 8

## Zero Field Residual Voltage

In most cases, the zero field residual voltage ( $V_{MT}$ ) temperature drift is separate, but not distinguishable, from sensitivity drift. When magnetic flux densities are low, the effect of ( $V_{MT}$ ) temperature drift of the residual voltage may be equal to or larger than the sensitivity change. The inability to consistently and accurately predict the rate and direction of drift in this offset voltage is the most difficult part of compensation.

Since the major portion of the  $V_{MT}$  is caused by unbalanced output contact po-

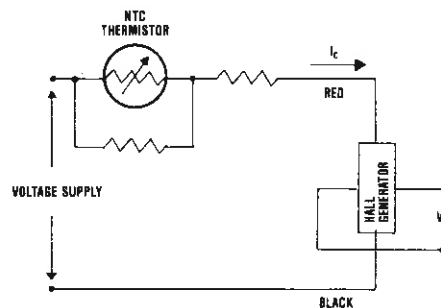
sitioning (misalignment voltage), the characteristics of the temperature drift may be predicted with fair accuracy. For devices with a room temperature  $V_{MT}$  greater than 500  $\mu$ V, the magnitude of the zero field residual voltage increases at a rate approximately equal to the temperature coefficient of resistance,  $\alpha_T$ . Units with  $V_{MT}$  between 100 and 500  $\mu$ V normally exhibit an increase in residual voltage magnitude, but at rates varying within  $\pm 100\%$  of the resistive coefficient. Below 100  $\mu$ V, the task of predicting the offset drift becomes difficult because the

The circuits described herein are intended only to illustrate typical applications. This information is believed reliable, however, no responsibility is assumed for inaccuracies.

# APPLICATION AIDS

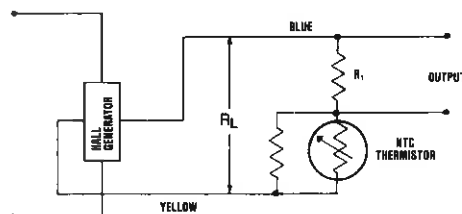
direction and rate are random. A circuit that can be used to compensate for the zero field residual voltage temperature drift is shown in Figure 12.

The thermistor shown compensates for a positive-going temperature drift (with respect to the output terminals). If zero field residual voltage varies negatively with increasing temperature, the thermistor network and  $R_1$  must be interchanged.



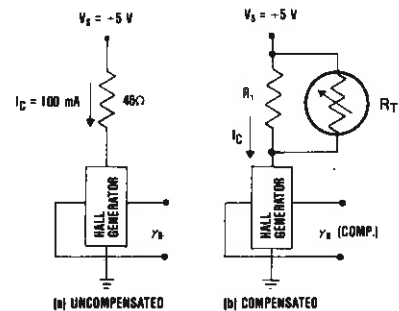
**Figure 9**

Hall generator can be compensated at its input by a negative temperature coefficient thermistor.



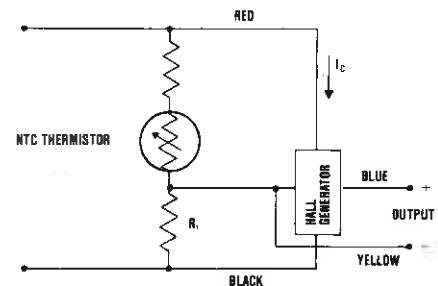
**Figure 10**

Temperature compensation of a Hall generator using a negative temperature coefficient thermistor.



**Figure 11**

The use of a negative temperature coefficient thermistor as a compensator in the Hall generator input.



**Figure 12**

Compensation of zero field voltage temperature drift of Hall generators.

H.Weiss. Structure and Application of Galvanomagnetic Devices, Pergamon Press  
H.H.Weider. Hall Generators and Magneto-resistors, Pion Limited  
MIL-STD-793-1 (WP), "Definitions Letter Symbols", Color Code and Circuit Code and Circuit Symbol for Devices Hall Effect

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